

# ECE 342

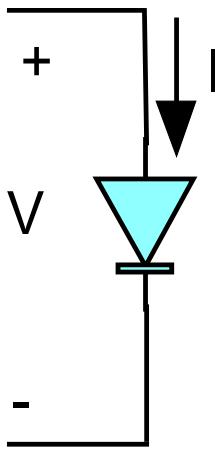
# Electronic Circuits

## Lecture 4

## Diode Models

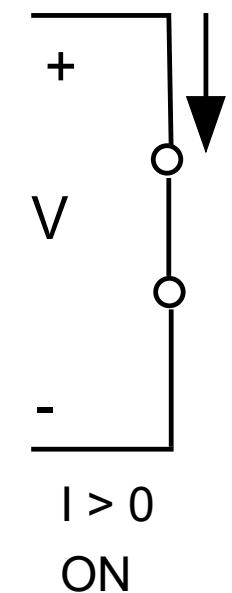
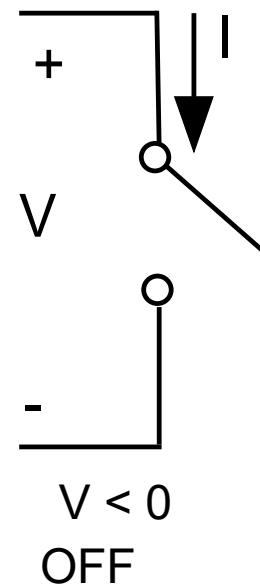
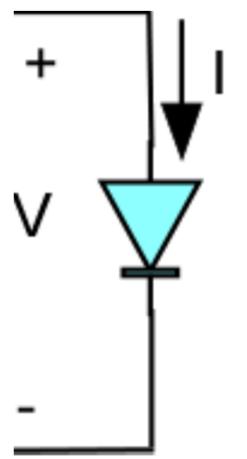
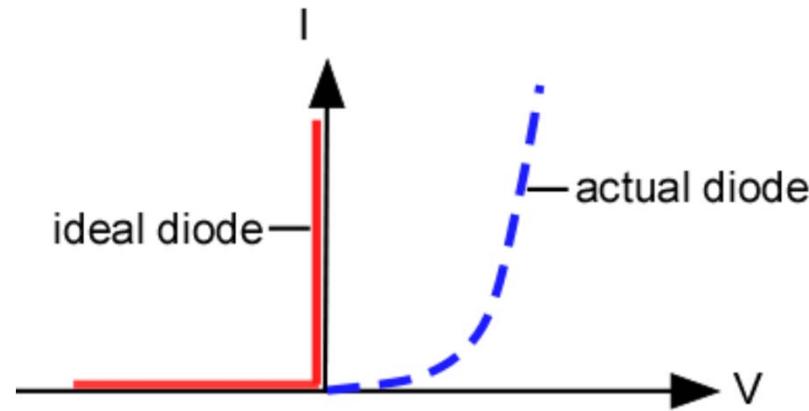
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# The Diode

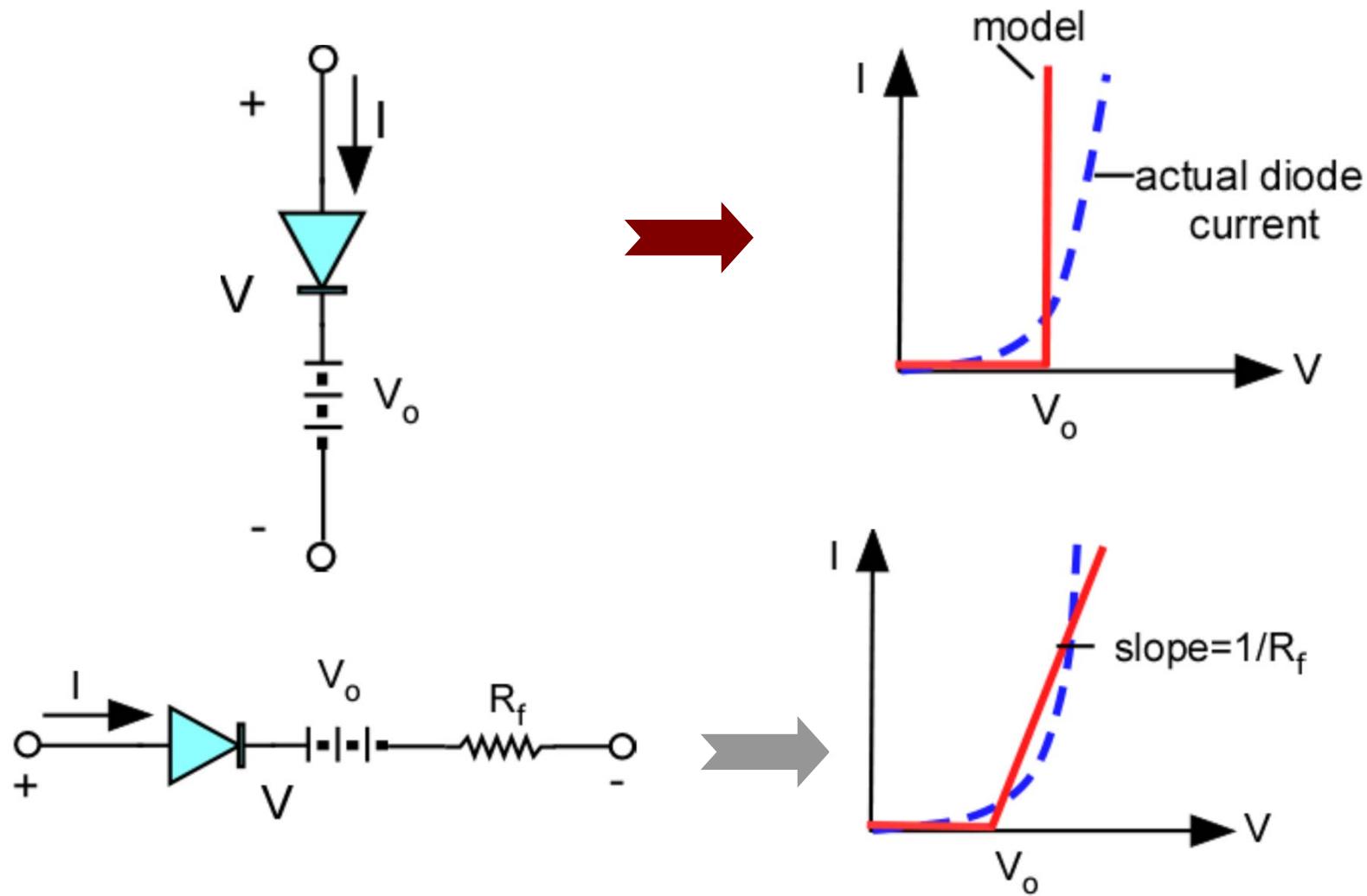


- **Diode Properties**
  - Two-terminal device that conducts current freely in one direction but blocks current flow in the opposite direction.
  - The two electrodes are the anode which must be connected to a positive voltage with respect to the other terminal, the cathode in order for current to flow.

# Ideal Diode Characteristics

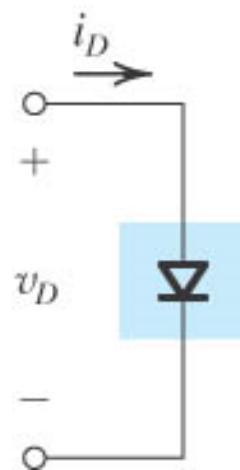
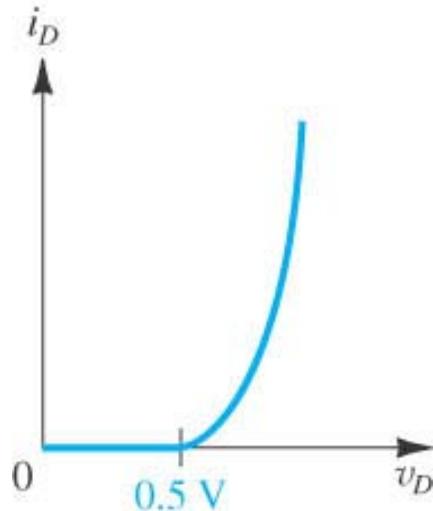


# Ideal Diode Characteristics

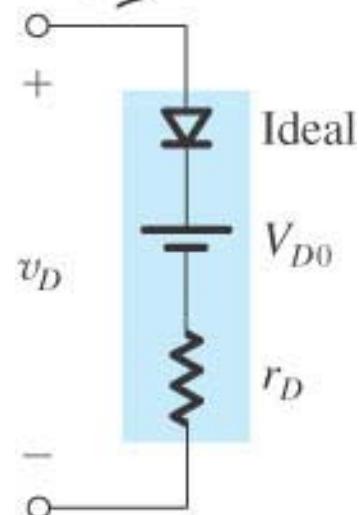
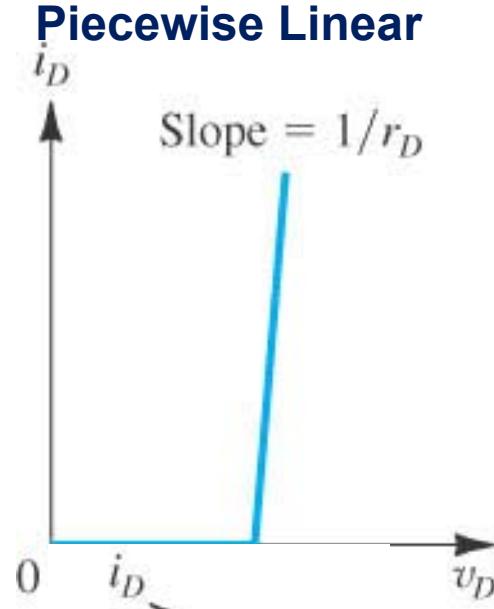


# Diode Models

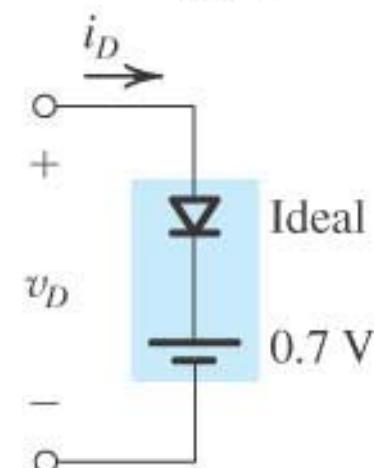
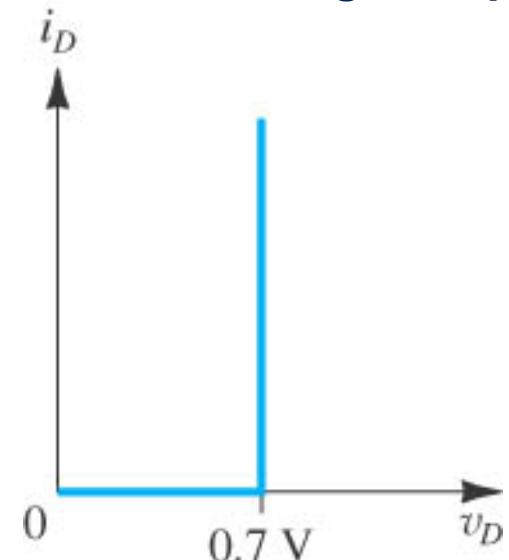
Exponential



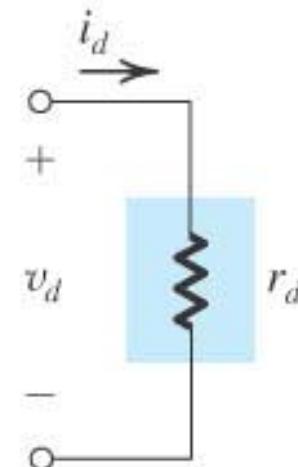
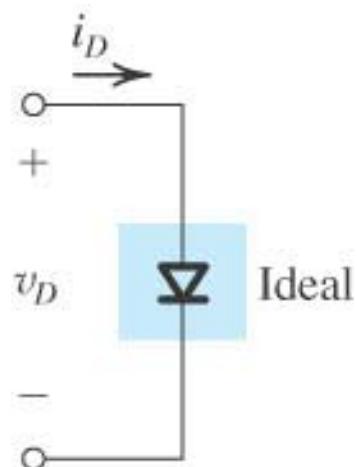
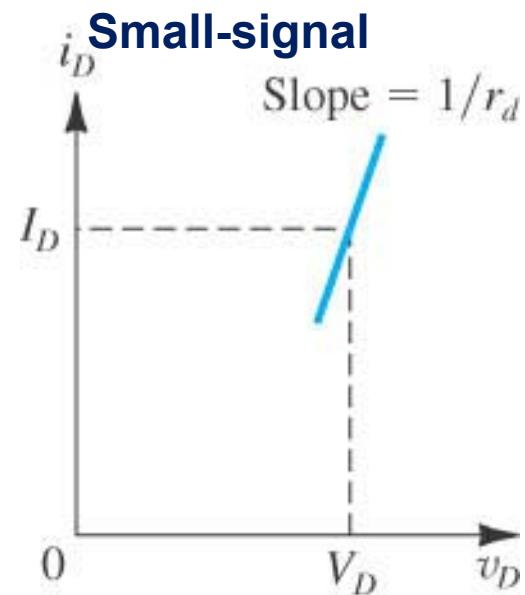
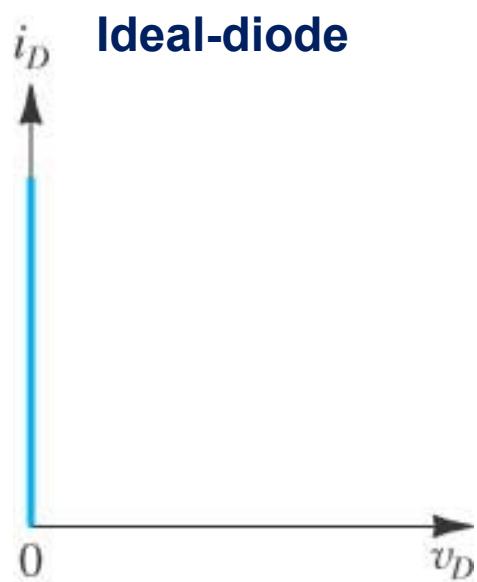
Piecewise Linear



Constant-Voltage-Drop



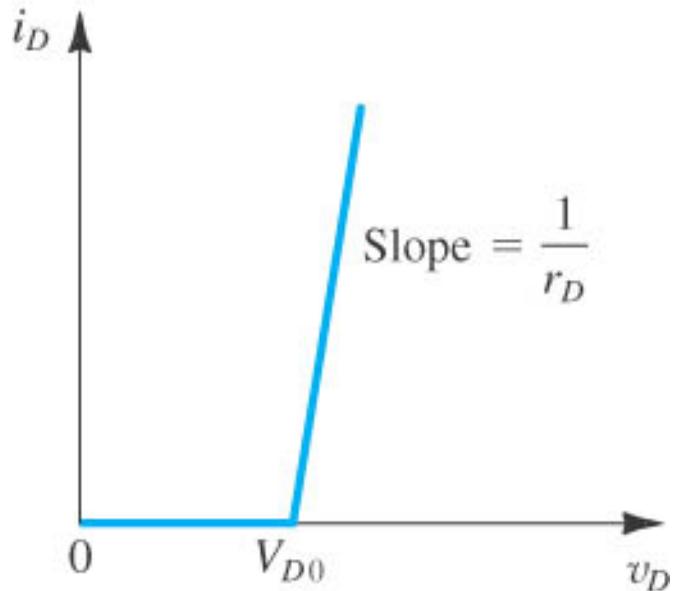
# Diode Models



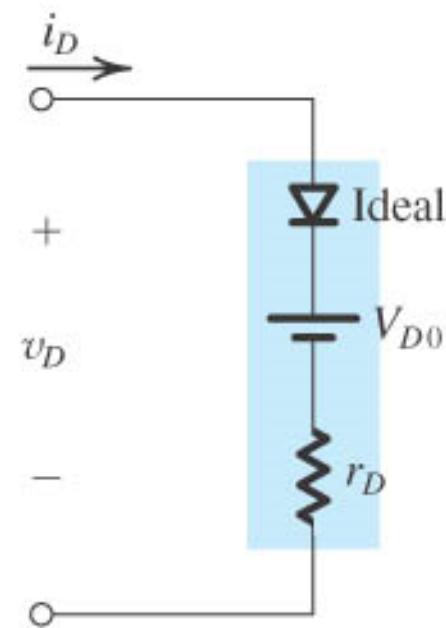
# Piecewise-Linear Model

for  $v_D \leq V_{D0}$  :  $i_D = 0$

for  $v_D \geq V_{D0}$  :  $i_D = \frac{1}{r_D} (v_D - V_{D0})$

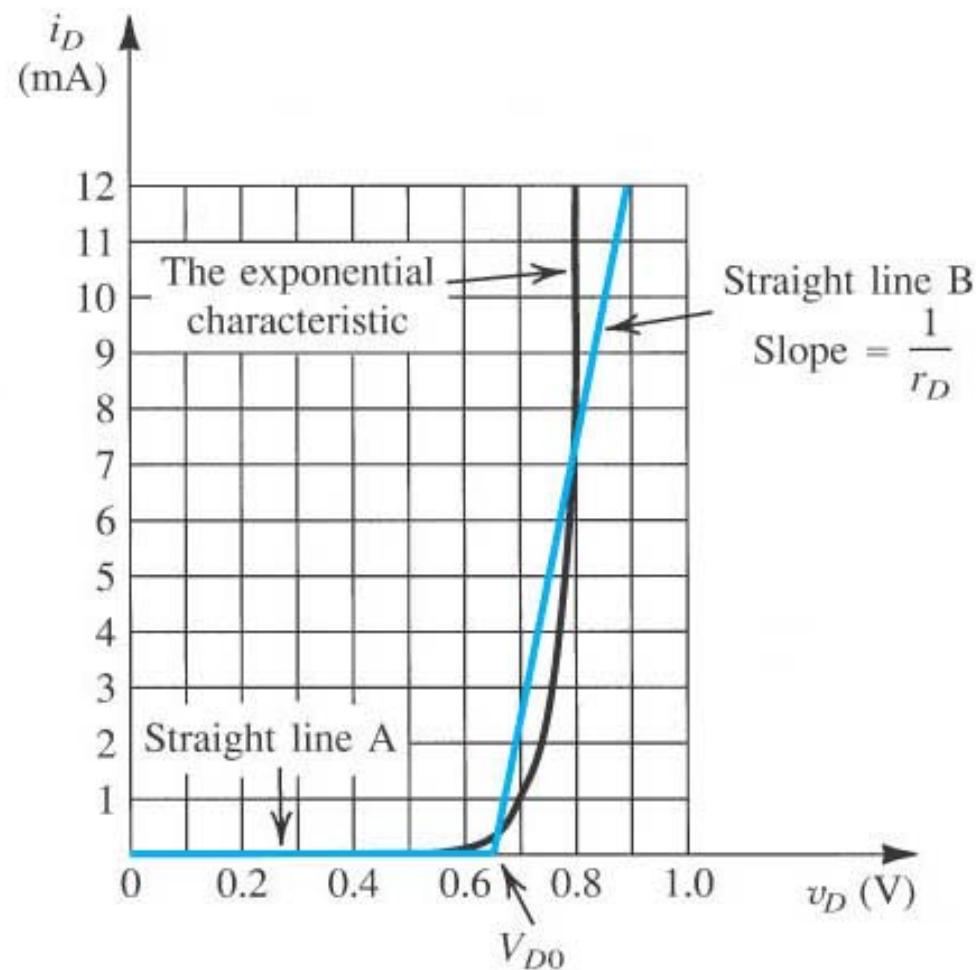


(a)



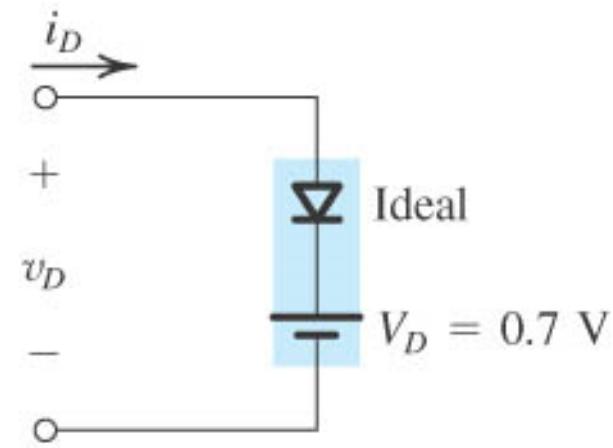
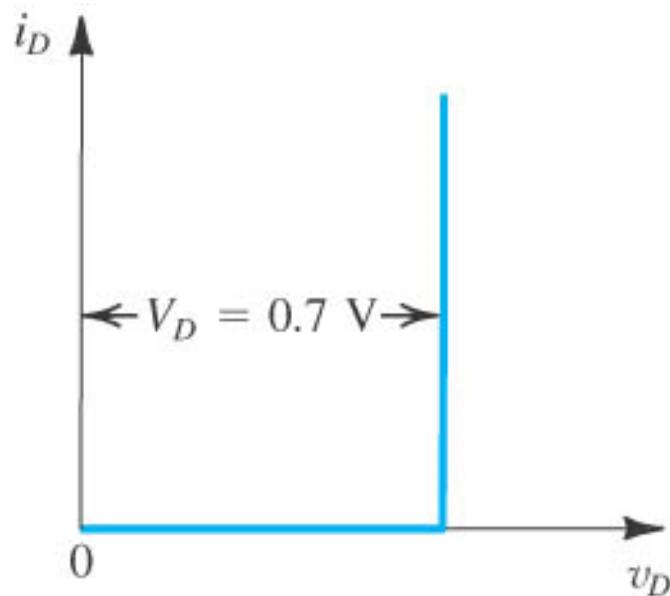
(b)

# Piecewise-Linear Model

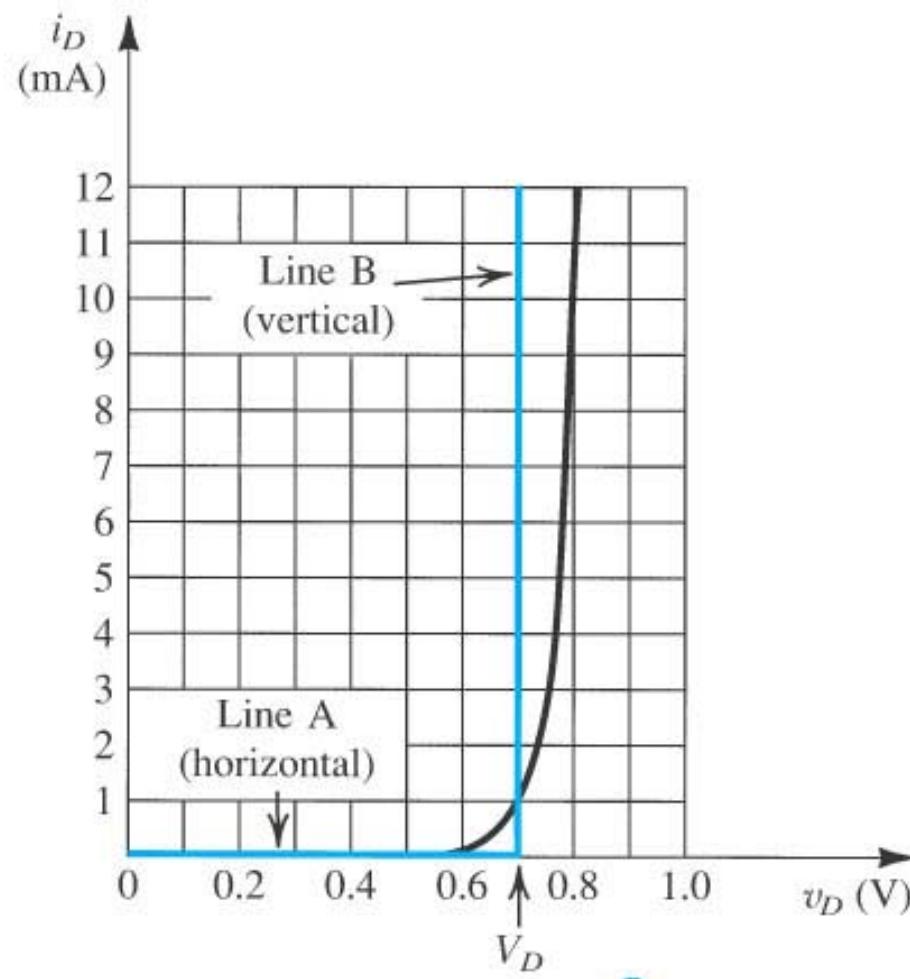


# Constant-Voltage-Drop Model

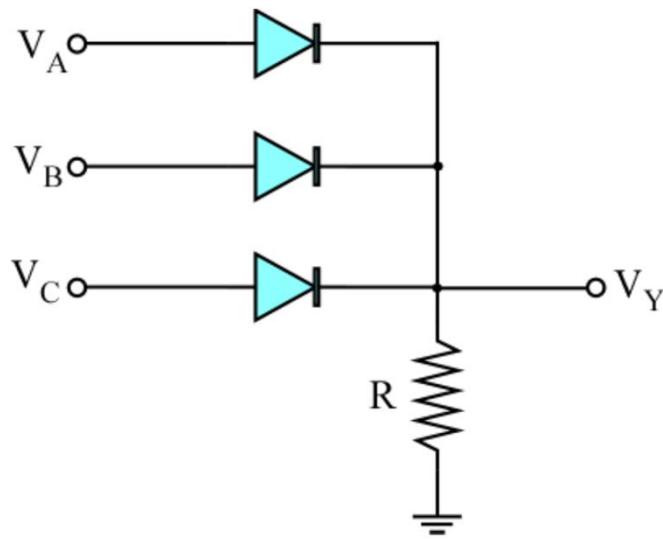
for  $i_D > 0$ :  $v_D = 0.7 \text{ V}$



# Constant-Voltage-Drop Model

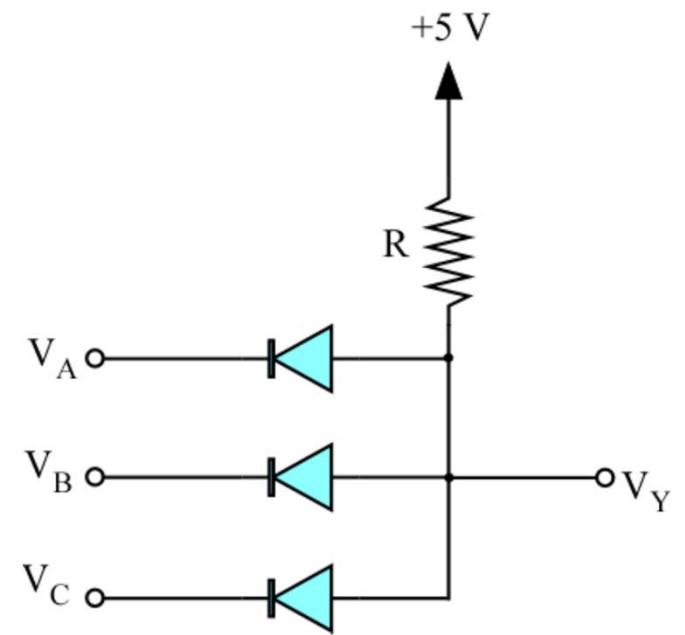


# Diodes Logic Gates



**OR Function**

$$Y = A + B + C$$



**AND Function**

$$Y = A \cdot B \cdot C$$

# Diode Circuit Example 1

**IDEAL Diodes**

Assume both diodes are on; then

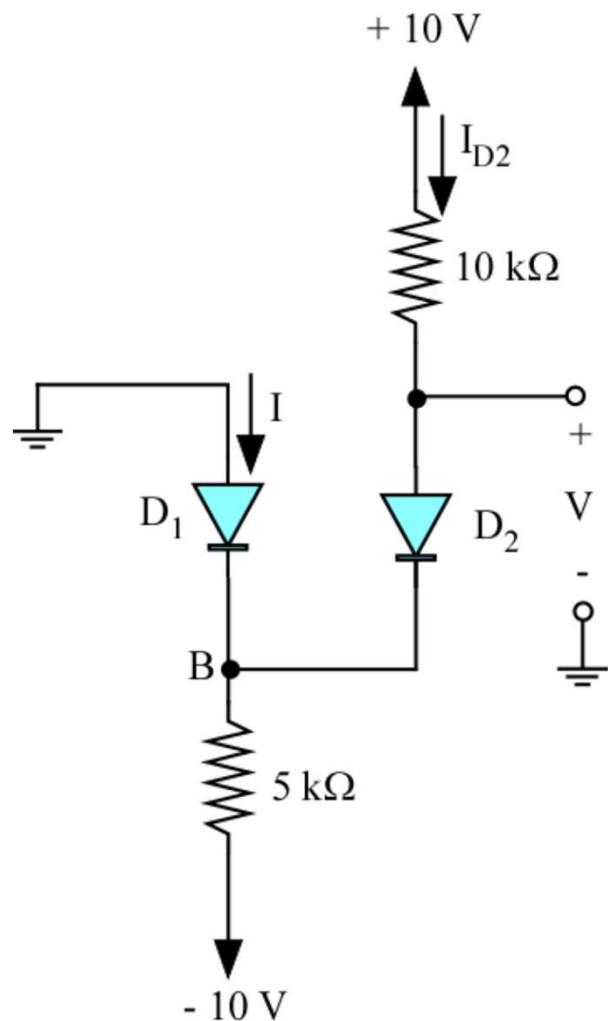
$$V_B = 0 \quad \text{and} \quad V = 0$$

$$I_{D2} = \frac{10 - 0}{10} = 1 \text{ mA}$$

**At node B**

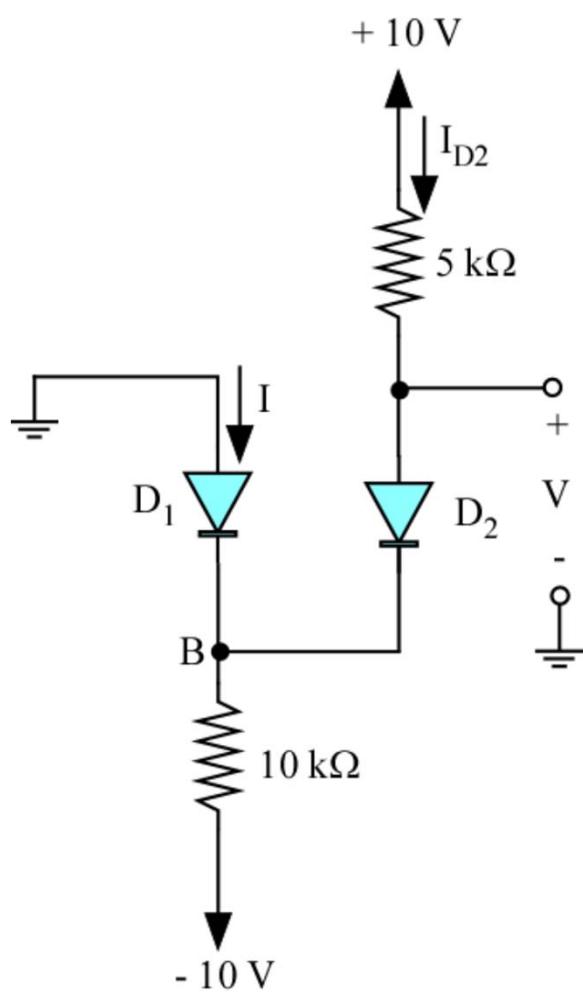
$$I + 1 = \frac{0 - (-10)}{5} \Rightarrow I = 1 \text{ mA}, V = 0 \text{ V}$$

**D<sub>1</sub> is conducting as originally assumed**



# Diode Circuit Example 2

**IDEAL Diodes**



Assume both diodes are on; then

$$V_B = 0 \quad \text{and} \quad V = 0$$

$$I_{D2} = \frac{10 - 0}{5} = 2 \text{ mA}$$

**At node B**

$$I + 2 = \frac{0 - (-10)}{10} \Rightarrow I = -1 \text{ mA} \Rightarrow \text{wrong}$$

**original assumption is not correct ...  
assume D<sub>1</sub> is off and D<sub>2</sub> is on**

$$I_{D2} = \frac{10 - (-10)}{15} = 1.33 \text{ mA}$$

$$V_B = -10 + 10 \times 1.33 = +3.3 \text{ V}$$

**D<sub>1</sub> is reverse biased as assumed**

# Example

The diode has a value of  $I_S = 10^{-12}$  mA at room temperature (300° K)

- (a) Approximate the current  $I$  assuming the voltage drop across the diode is 0.7V
- (b) Calculate the accurate value of  $I$
- (c) If  $I_S$  doubles for every 6° C increase in temperature, repeat part (b) if the temperature increases by 40° C

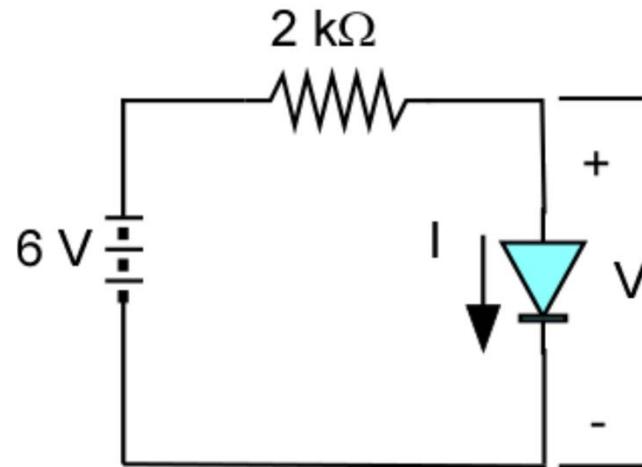
(a) The resistor will have an approximate voltage of  $6 - 0.7 = 5.3$  V. Ohm's law then gives a current of

$$I = \frac{5.3}{2} = 2.65 \text{ mA}$$

(b) The current through the resistor must equal the diode current; so we have

$$I = \frac{6 - V}{2} \text{ (resistor current)}$$

$$I = I_S e^{V/V_T} \text{ (diode current)}$$



# Example (cont'd)

$$\frac{6-V}{2} = 10^{-12} e^{V/0.026}$$

Nonlinear equation → must be solved iteratively

Solution:  $V = 0.744$  V

Using this value of the voltage, we can calculate the current

$$I = \frac{6-V}{2} = \frac{6-0.744}{2} = 2.63 \text{ mA}$$

When the temperature changes, both  $I_s$  and  $V_T$  will change. Since  $V_T = kT/q$  varies directly with  $T$ , the new value is:

$$V_T(340) = V_T(300) \times \frac{340}{300} = 0.0295$$

# Example (cont'd)

The value of  $I_s$  doubles for each  $6^\circ C$  increase, thus the new value of  $I_s$  is

$$I_s(340) = I_s(300) \times 2^{40/6} = 1.016 \times 10^{-10} \text{ mA}$$

The equation for  $I$  is then

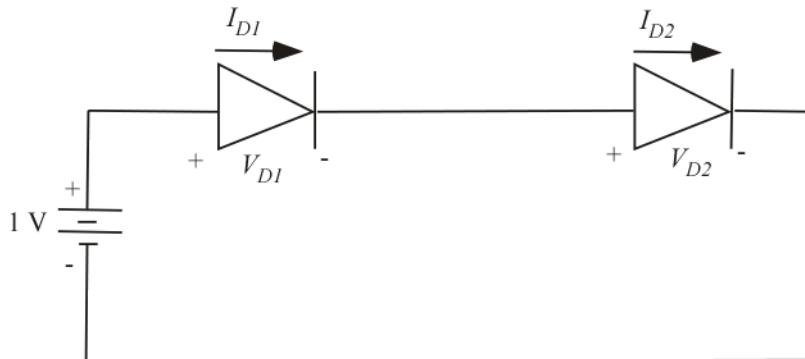
$$I = \frac{6 - V}{2} = 1.016 \times 10^{-10} \times e^{V/0.0295}$$

Solving iteratively, we get

$$V = 0.640 \text{ V} \quad \text{and} \quad I = 2.68 \text{ mA}$$

# Example

Two diodes are connected in series as shown in the figure with  $I_{s1} = 10^{-16}$  A and  $I_{s2} = 10^{-14}$  A. If the applied voltage is 1 V, calculate the currents  $I_{D1}$  and  $I_{D2}$  and the voltage across each diode  $V_{D1}$  and  $V_{D2}$ .



The diode equations can be written as:

$$I_{D1} = I_{s1} e^{V_{D1}/V_T} \quad I_{D2} = I_{s2} e^{V_{D2}/V_T} \quad \frac{I_{s1}}{I_{s2}} e^{\frac{V_{D1}-V_{D2}}{V_T}} = \frac{I_{D1}}{I_{D2}} = 1$$

from which  $V_{D1} - V_{D2} = -V_T \ln\left(\frac{I_{s1}}{I_{s2}}\right) = -0.12$

Using KVL, we get  $V_{D1} + V_{D2} = 1$  from which  $V_{D2} = 0.44$  V and  $V_{D1} = 0.56$  V

$$I_{D1} = 10^{-16} e^{0.56/0.026} = 0.22 \mu\text{A} = I_{D2}$$

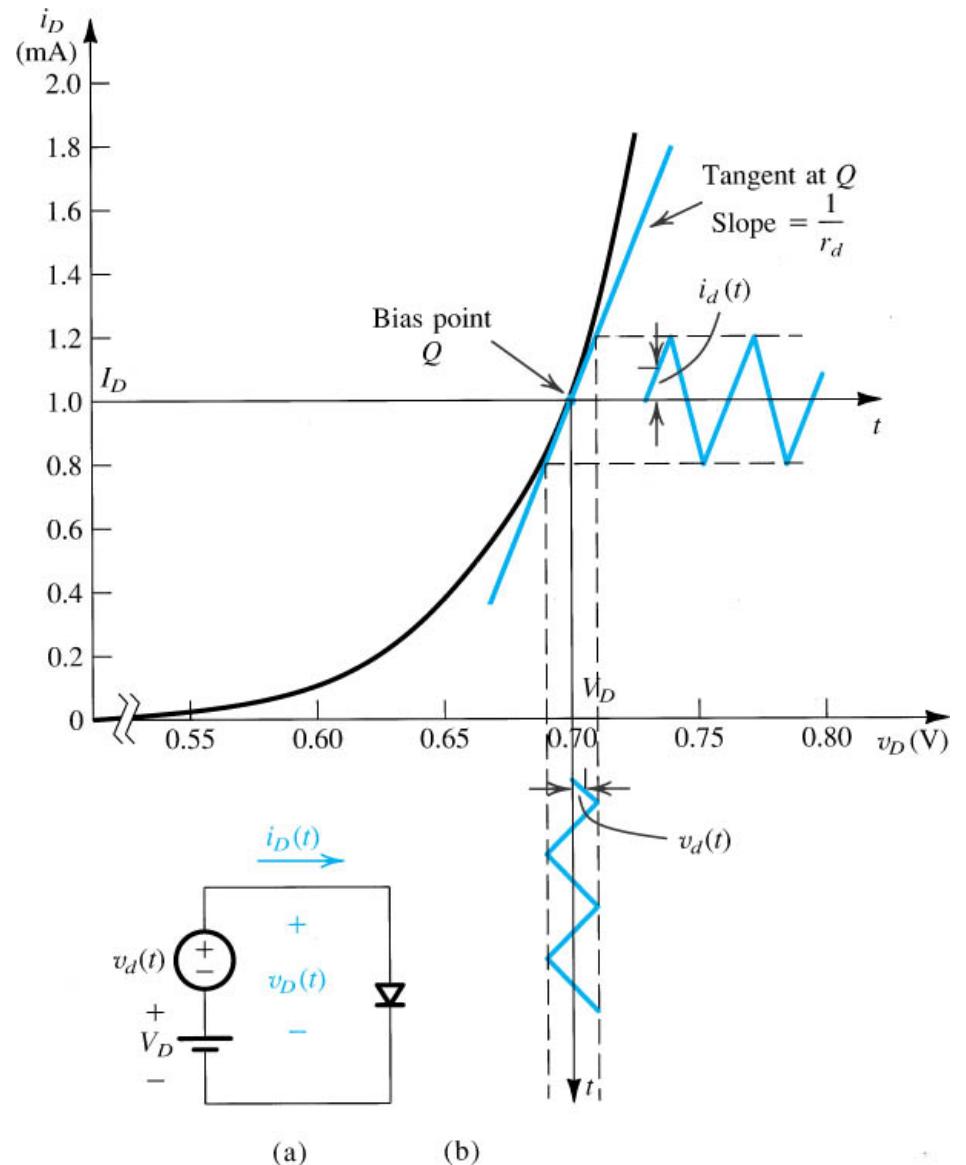
# Small Signal Model

**Approximation - valid for small fluctuations about bias point**

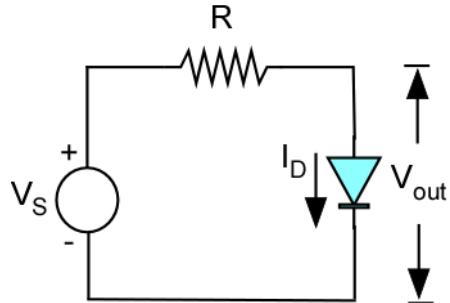
$$i_D = I_D e^{v_d / nV_T}$$

$$r_d = \frac{1}{\left[ \frac{\partial i_D}{\partial v_D} \right]_{i_D=I_D}} = \frac{nV_T}{I_D}$$

$$\begin{aligned} i_D &= I_D + i_d \\ \text{Total} &\quad \text{DC} \quad \text{applied (small)} \\ v_D &= V_D + v_d \end{aligned}$$



# Diode Circuits

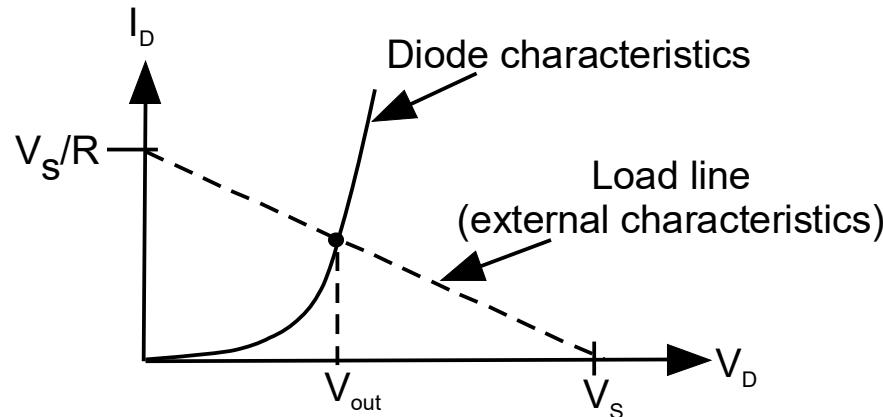


$$V_{out} = V_D$$

$$I_D = I_S \left( e^{V_D/V_T} - 1 \right)$$

$$V_s = RI_D + V_D = RI_D(V_D) + V_D$$

**Nonlinear transcendental system → Use graphical method**



**Solution is found at intersection of load line characteristics and diode characteristics**

# Diode Circuits – Iterative Methods

## Newton-Raphson Method

Wish to solve  $f(x)=0$  for  $x$

$$\text{Use: } x_{k+1} = x_k - [f'(x_k)]^{-1} f(x_k)$$

$$x^{(k+1)} = x^{(k)} - [f'(x^{(k)})]^{-1} f(x^{(k)})$$

$$f(V_D) = \frac{V_D - V_S}{R} + I_S \left( e^{V_D/V_T} - 1 \right) = 0$$
$$f'(V_D) = \frac{1}{R} + \frac{I_S}{V_T} e^{V_D/V_T}$$

$$V_D^{(k+1)} = V_D^{(k)} - \frac{\frac{V_D^{(k)} - V_S}{R} + I_S \left( e^{V_D^{(k)}/V_T} - 1 \right)}{\frac{1}{R} + \frac{I_S}{V_T} e^{V_D^{(k)}/V_T}}$$

Where  $V_D^{(k)}$  is the value of  $V_D$  at the  $k$ th iteration

Procedure is repeated until convergence to final (true) value of  $V_D$  which is the solution. Rate of convergence is quadratic.

