

ECE 546

Lecture 02

Review of Electromagnetics

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Jose E. Schutt-Aine
Electrical & Computer Engineering
University of Illinois
jesa@illinois.edu

Electromagnetic Quantities

$\vec{E}$ Electric field (Volts/m)

$\vec{D}$ Electric flux density (Coulombs/m²)

$\vec{H}$ Magnetic field (Amperes/m)

$\vec{B}$ Magnetic flux density (Webers/m²)

$\vec{J}$ Current density (Amperes/m²)

ρ Charge density (Coulombs/m²)

Maxwell's Equations

$$\nabla \times \vec{E} = -\frac{\partial \vec{B}}{\partial t}$$

Faraday's Law of Induction

$$\nabla \times \vec{H} = \vec{J} + \frac{\partial \vec{D}}{\partial t}$$

Ampère's Law

$$\nabla \cdot \vec{D} = \rho$$

Gauss' Law for electric field

$$\nabla \cdot \vec{B} = 0$$

Gauss' Law for magnetic field

Constitutive Relations

$$\vec{B} = \mu \vec{H} \qquad \vec{D} = \varepsilon \vec{E}$$

Permittivity ε : Farads/m

Permeability μ : Henries/m

Free Space

$$\varepsilon_o = 8.85 \times 10^{-12} \text{ } F / m$$

$$\mu_o = 4\pi \times 10^{-7} \text{ } H / m$$

Continuity Equation

$$\nabla \times \vec{H} = \vec{J} + \frac{\partial \vec{D}}{\partial t}$$

$$\nabla \cdot \nabla \times \vec{H} = \nabla \cdot \vec{J} + \nabla \cdot \frac{\partial \vec{D}}{\partial t} = \nabla \cdot \vec{J} + \frac{\partial}{\partial t} \nabla \cdot \vec{D} = 0$$

$$\nabla \cdot \vec{D} = \rho$$

$$\nabla \cdot \vec{J} + \frac{\partial \rho}{\partial t} = 0$$

Electrostatics

Assume no time dependence $\frac{\partial}{\partial t} \rightarrow 0$

$$\nabla \times \vec{E} = 0$$

Since $\nabla \times \vec{E} = 0$, $\exists \phi$ such that

$$\vec{E} = -\nabla \phi \quad \text{where } \phi \text{ is the scalar potential}$$

$$\text{with } \nabla \cdot \vec{E} = \frac{\rho}{\epsilon} \quad \text{we get } \nabla \cdot (-\nabla \phi) = -\nabla^2 \phi = \frac{\rho}{\epsilon}$$

$$\text{we get } \nabla \cdot (-\nabla \phi) = -\nabla^2 \phi = \frac{\rho}{\epsilon} \quad \leftarrow \text{Poisson's Equation}$$

if no charge is present $\nabla^2 \phi = 0 \quad \leftarrow \text{Laplace's Equation}$

Integral Form of ME

$$\oint_C \vec{E} \cdot d\vec{l} = -\frac{\partial}{\partial t} \int_S \vec{B} \cdot d\vec{S}$$

Faraday's Law of Induction

$$\oint_C \vec{H} \cdot d\vec{l} = \int_S \vec{J} \cdot d\vec{S} + \frac{\partial}{\partial t} \int_S \vec{D} \cdot d\vec{S}$$

Ampere's Law

$$\oint_S \vec{D} \cdot d\vec{S} = \int_V \rho \cdot dv$$

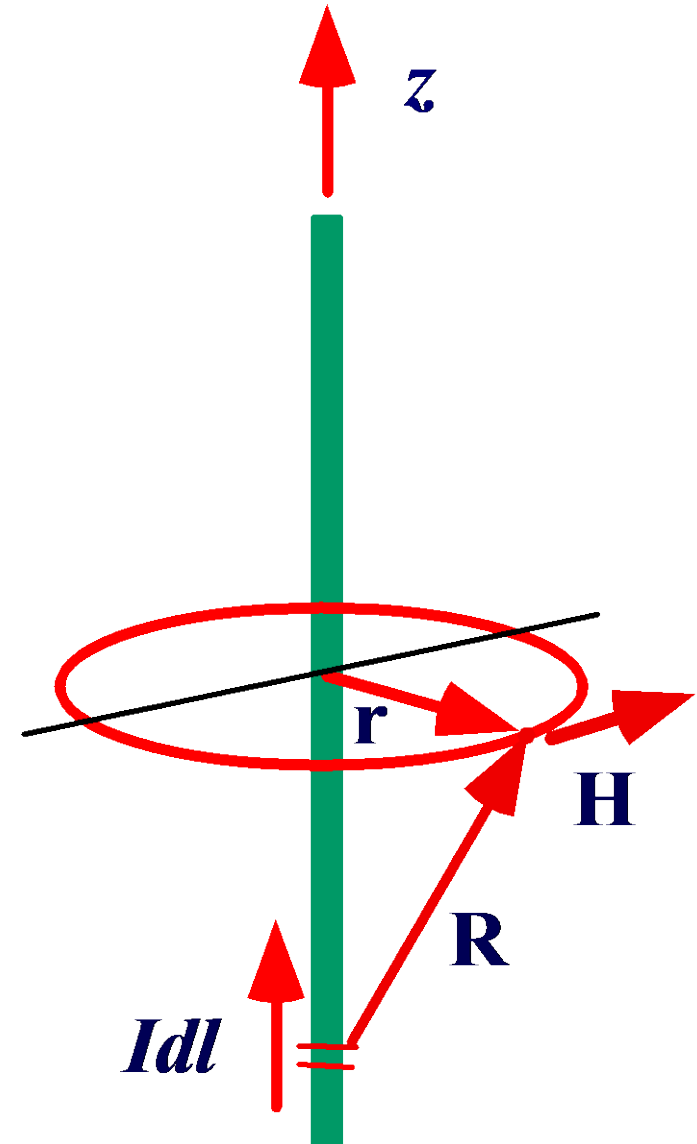
Gauss' Law – D vector

$$\oint_S \vec{B} \cdot d\vec{S} = 0$$

Gauss' Law – B Field

H-Field – Infinite Length Wire

- Use azimuthal symmetry
- Assume return path is at infinity
- Use Stokes' theorem



Using Gauss' Law – D Field

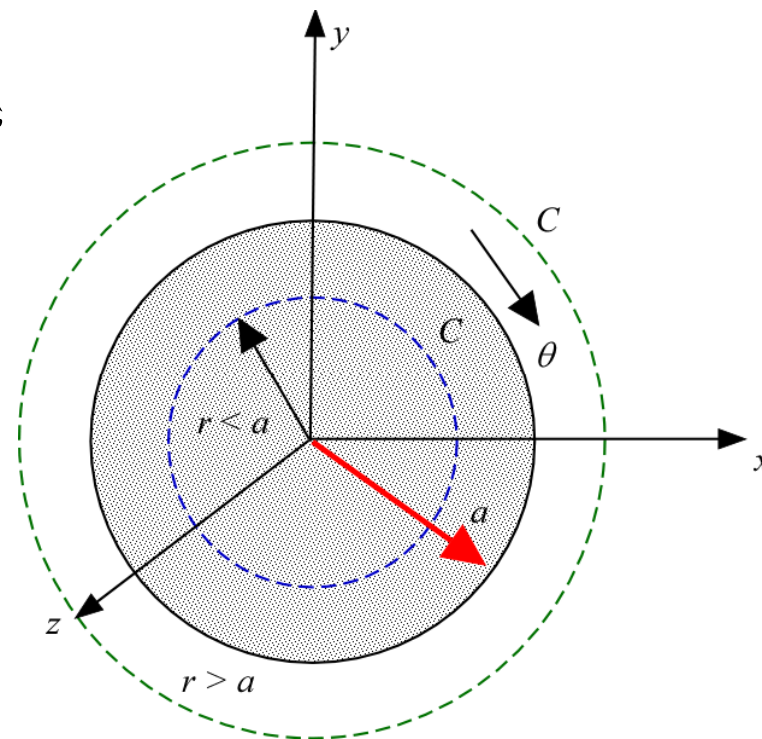
Spherical volume of radius a with uniform charge density $\rho_o(\text{C/m}^3)$

$$\oint_S \vec{D} \cdot d\vec{S} = \int_V \rho \cdot dv$$

$$\oint_S \vec{D} \cdot d\vec{S} = \int_{\theta=0}^{\pi} \int_{\phi=0}^{2\pi} D_r(r) \hat{r} \cdot r^2 \sin \theta d\theta d\phi \hat{r}$$

Charge exists only for $r < a$ and with uniform density

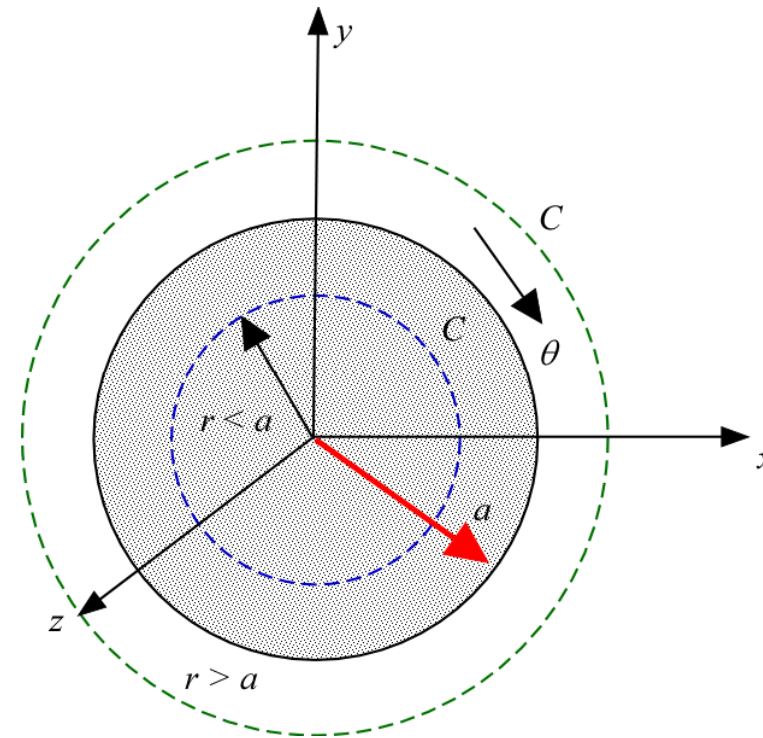
$$\int_V \rho \cdot dv = \begin{cases} \frac{4}{3} \pi r^3 \rho_o & \text{for } r \leq a \\ \frac{4}{3} \pi a^3 \rho_o & \text{for } r \geq a \end{cases}$$



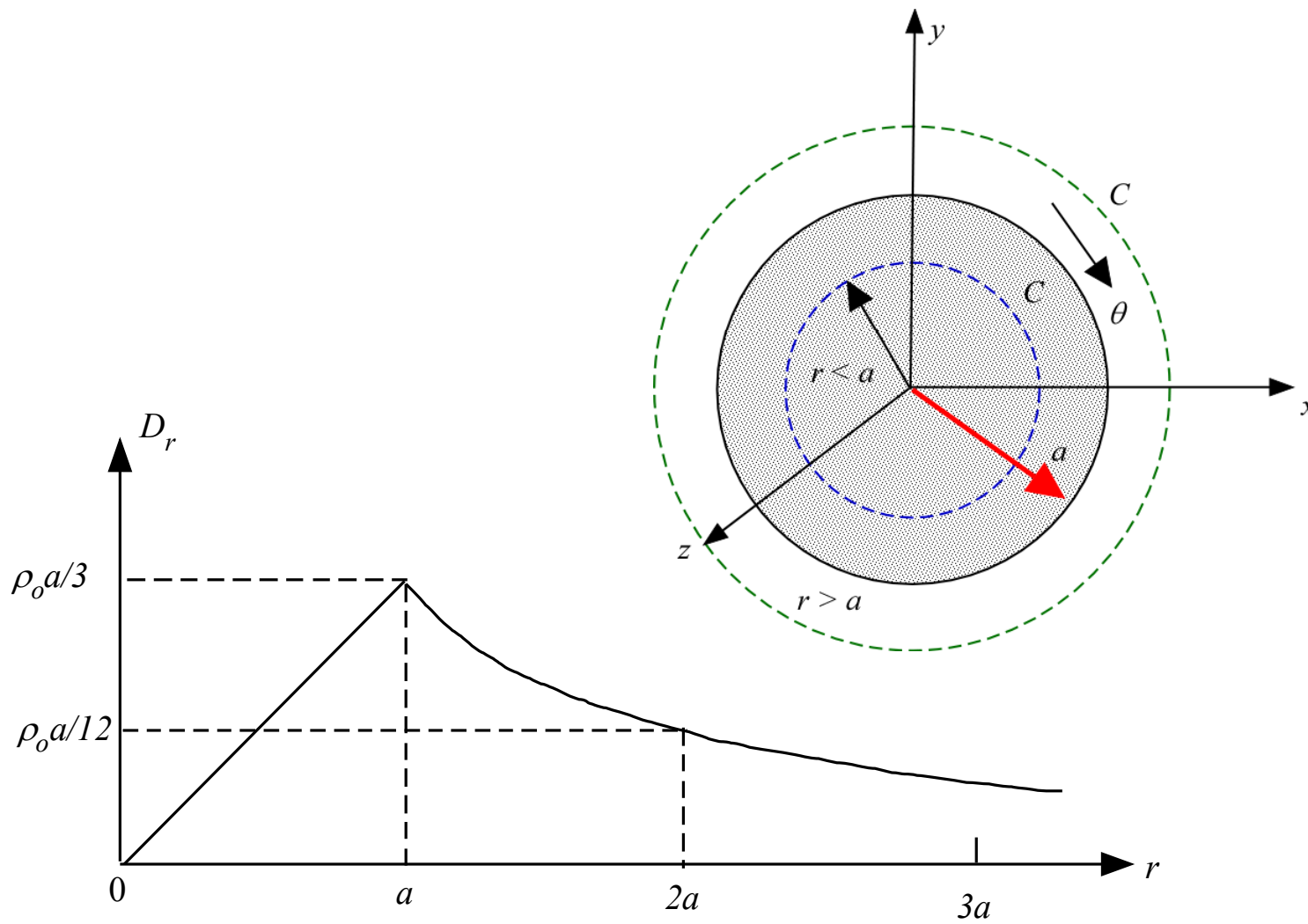
Using Gauss' Law – D Field

$$4\pi r^2 D_r(r) = \begin{cases} \frac{4}{3}\pi r^3 \rho_o & \text{for } r \leq a \\ \frac{4}{3}\pi a^3 \rho_o & \text{for } r \geq a \end{cases}$$

$$D_r(r) = \begin{cases} \frac{\rho_o r}{3} & \text{for } r \leq a \\ \frac{\rho_o a^3}{3r^2} & \text{for } r \geq a \end{cases}$$

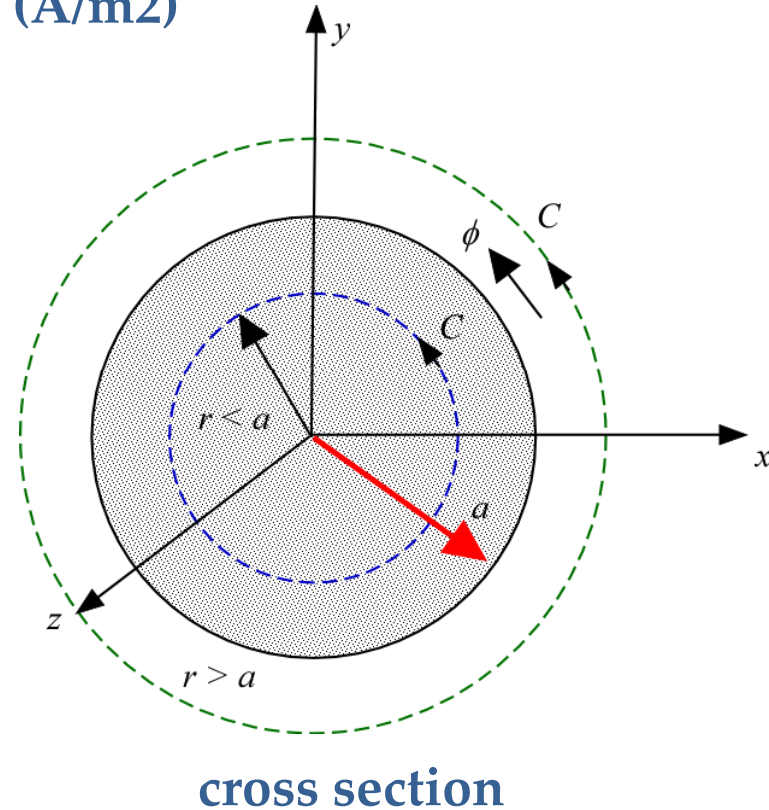
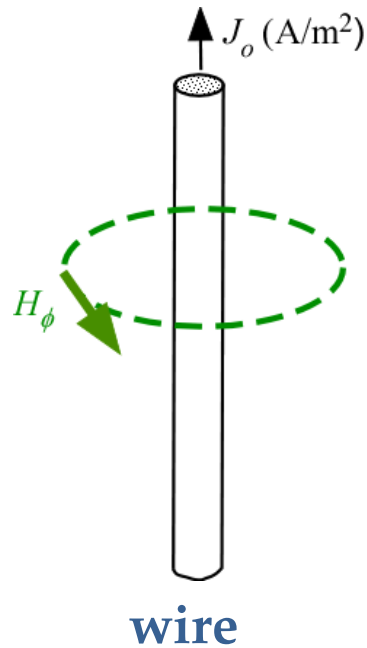


Using Gauss' Law – D Field



Using Gauss' Law – H Field

Infinitely long cylindrical wire of radius a
with uniform current density J_o (A/m²)



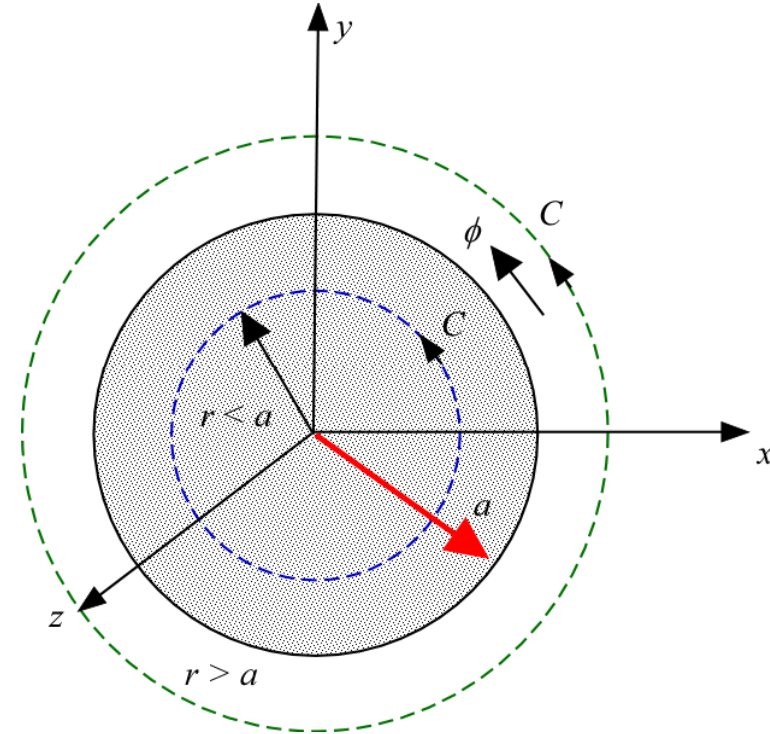
Need to calculate H field both inside and outside wire →
Use azimuthal symmetry

Using Gauss' Law – H Field

$$\oint_C \vec{H} \cdot d\vec{l} = \int_S \vec{J} \cdot d\vec{S} + \frac{\partial}{\partial t} \int_S \vec{D} \cdot d\vec{S}$$

$$\oint_C \vec{H} \cdot d\vec{l} = \int_{\phi=0}^{2\pi} H_{\phi}(r) \hat{\phi} \cdot r d\phi \hat{\phi} = 2\pi r H_{\phi}(r)$$

Current exists only for $r < a$ and with uniform density

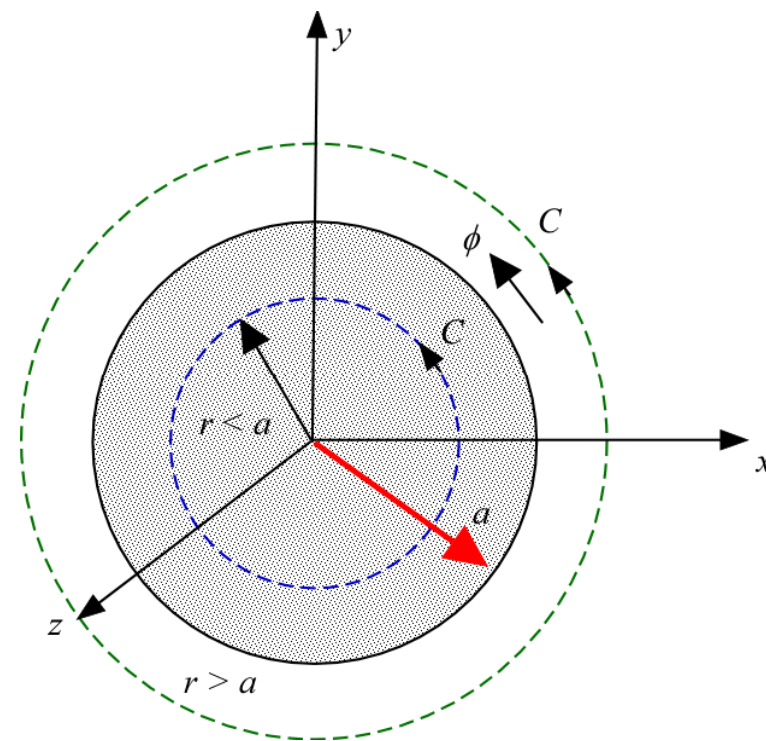


$$\int_S \vec{J} \cdot d\vec{S} = \begin{cases} \int_{r=0}^r \int_{\phi=0}^{2\pi} J_o \hat{z} \cdot r dr d\phi \hat{z} & \text{for } r \leq a \\ \int_{r=0}^a \int_{\phi=0}^{2\pi} J_o \hat{z} \cdot r dr d\phi \hat{z} & \text{for } r \geq a \end{cases}$$

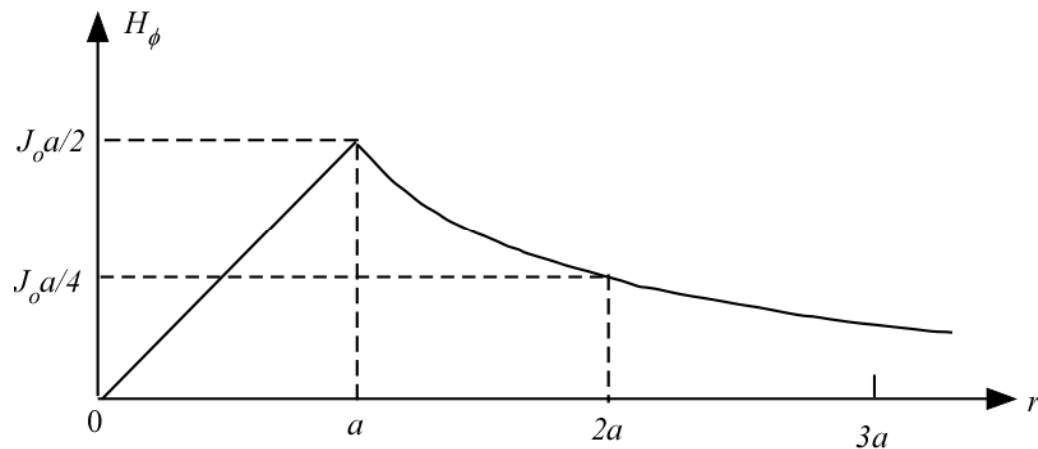
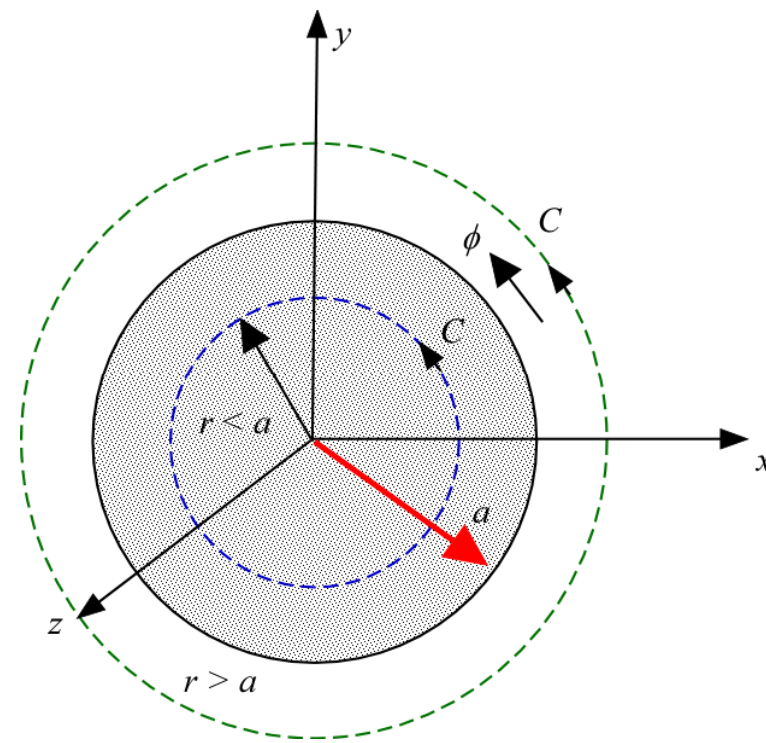
Using Gauss' Law – H Field

$$2\pi r H_{\phi} = \begin{cases} J_o \pi r^2 & \text{for } r \leq a \\ J_o \pi a^2 & \text{for } r \geq a \end{cases}$$

$$H_{\phi} = \begin{cases} \frac{J_o r}{2} & \text{for } r \leq a \\ \frac{J_o a^2}{2r} & \text{for } r \geq a \end{cases}$$



Using Gauss' Law – H Field



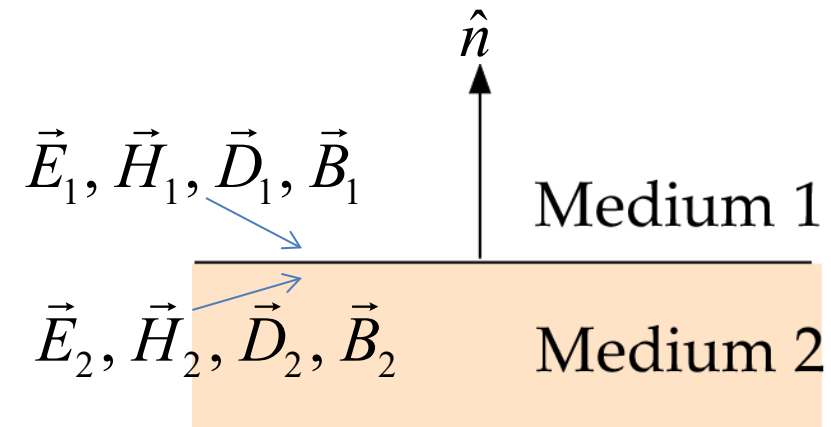
Boundary Conditions

$$\hat{n} \times (\vec{E}_1 - \vec{E}_2) = 0$$

$$\hat{n} \times (\vec{H}_1 - \vec{H}_2) = \vec{J}_S$$

$$\hat{n} \cdot (\vec{D}_1 - \vec{D}_2) = \rho_S$$

$$\hat{n} \cdot (\vec{B}_1 - \vec{B}_2) = 0$$



Maxwell's Equations in Free Space

$$\nabla \times \vec{E} = -\mu_o \frac{\partial \vec{H}}{\partial t} \quad \text{Faraday's Law of Induction}$$

$$\nabla \times \vec{H} = \varepsilon_o \frac{\partial \vec{E}}{\partial t} \quad \text{Ampère's Law}$$

$$\nabla \cdot \vec{E} = 0 \quad \text{Gauss' Law for electric field}$$

$$\nabla \cdot \vec{B} = 0 \quad \text{Gauss' Law for magnetic field}$$

Wave Equation

$$\nabla \times (\nabla \times \vec{E}) = -\mu_o \frac{\partial}{\partial t} (\nabla \times \vec{H})$$

$$\nabla (\nabla \cdot \vec{E}) - \nabla^2 \vec{E} = -\mu_o \frac{\partial}{\partial t} \epsilon_o \frac{\partial \vec{E}}{\partial t}$$

$$\nabla^2 \vec{E} = \mu_o \epsilon_o \frac{\partial^2 \vec{E}}{\partial t^2} \quad \leftarrow \text{Wave Equation}$$

can show that

$$\nabla^2 \vec{H} = \mu_o \epsilon_o \frac{\partial^2 \vec{H}}{\partial t^2}$$

Wave Equation

$$\frac{\partial^2 \vec{E}}{\partial x^2} + \frac{\partial^2 \vec{E}}{\partial y^2} + \frac{\partial^2 \vec{E}}{\partial z^2} = \mu_o \epsilon_o \frac{\partial^2 \vec{E}}{\partial t^2}$$

separating the components

$$\frac{\partial^2 E_x}{\partial x^2} + \frac{\partial^2 E_x}{\partial y^2} + \frac{\partial^2 E_x}{\partial z^2} = \mu_o \epsilon_o \frac{\partial^2 E_x}{\partial t^2}$$

$$\frac{\partial^2 E_y}{\partial x^2} + \frac{\partial^2 E_y}{\partial y^2} + \frac{\partial^2 E_y}{\partial z^2} = \mu_o \epsilon_o \frac{\partial^2 E_y}{\partial t^2}$$

$$\frac{\partial^2 E_z}{\partial x^2} + \frac{\partial^2 E_z}{\partial y^2} + \frac{\partial^2 E_z}{\partial z^2} = \mu_o \epsilon_o \frac{\partial^2 E_z}{\partial t^2}$$

Wave Equation → Plane Wave

(a) Assume that only E_x exists → $E_y=E_z=0$

(b) Only z spatial dependence → $\frac{\partial}{\partial x} = \frac{\partial}{\partial y} = 0$

$$\frac{\partial^2 E_x}{\partial z^2} = \mu_o \epsilon_o \frac{\partial^2 E_x}{\partial t^2}$$

This situation leads to the plane wave solution

In addition, assume a time-harmonic dependence

$$E_x \sim e^{j\omega t} \quad \text{then} \quad \frac{\partial}{\partial t} \rightarrow j\omega$$

Plane Wave Solution

$$\frac{\partial^2 E_x}{\partial z^2} = -\mu_o \epsilon_o \omega^2 E_x$$

solution $\begin{cases} E_x = E_{o+} e^{-j\beta z} & \text{forward wave} \\ E_x = E_{o-} e^{+j\beta z} & \text{backward wave} \end{cases}$

where $\beta = \omega \sqrt{\mu_o \epsilon_o}$ **propagation constant**

In the time domain

solution $\begin{cases} E_x(t) = E_{o+} \cos(\omega t - \beta z) & \text{forward traveling wave} \\ E_x(t) = E_{o-} \cos(\omega t + \beta z) & \text{backward traveling wave} \end{cases}$

Plane Wave Characteristics

where $\beta = \omega\sqrt{\mu_o\epsilon_o}$ propagation constant

$$\frac{\omega}{\beta} = \frac{1}{\sqrt{\mu\epsilon}} = v = \text{propagation velocity}$$

In free space $v = c = \frac{1}{\sqrt{\mu_o\epsilon_o}} = 3 \times 10^8 \text{ m / s}$

Solution for Magnetic Field

$$\nabla \times \vec{E} = -\mu \frac{\partial \vec{H}}{\partial t} \Rightarrow \nabla \times \vec{E} = -j\omega\mu \vec{H}$$

$$\vec{H} = -\frac{1}{j\omega\mu} \nabla \times \vec{E} = -\frac{1}{j\omega\mu} \begin{vmatrix} \hat{x} & \hat{y} & \hat{z} \\ \frac{\partial}{\partial x} & \frac{\partial}{\partial y} & \frac{\partial}{\partial z} \\ E_x & 0 & 0 \end{vmatrix}$$

If we assume that $\vec{E} = \hat{x}E_{o+}e^{-j\beta z}$ then $\vec{H} = \hat{y}\frac{E_{o+}}{\eta}e^{-j\beta z}$

If we assume that $\vec{E} = \hat{x}E_{o-}e^{+j\beta z}$ then $\vec{H} = -\hat{y}\frac{E_{o-}}{\eta}e^{+j\beta z}$

$\eta = \sqrt{\frac{\mu}{\epsilon}}$ intrinsic impedance of medium

Time-Average Poynting Vector

$$\vec{P}(t) = \vec{E}(t) \times \vec{H}(t) \quad \text{Poynting vector W/m}^2$$

time-average Poynting vector W/m²

$$\langle \vec{P} \rangle = \frac{1}{T} \int_0^T \vec{P}(t) dt = \frac{1}{T} \int_0^T \vec{E}(t) \times \vec{H}(t) dt$$

We can show that

$$\langle \vec{P} \rangle = \frac{1}{2} \text{Re} \left[\vec{E} \times \vec{H}^* \right]$$

where $\vec{E}$ and $\vec{H}$ are the phasors of $\vec{E}(t)$ and $\vec{H}(t)$ respectively

Material Medium

$$\nabla \times \vec{E} = -\frac{\partial \mu \vec{H}}{\partial t}$$

or

$$\nabla \times \vec{E} = -j\omega\mu\vec{H}$$

$$\nabla \times \vec{H} = \vec{J} + \frac{\partial \varepsilon \vec{E}}{\partial t}$$

$$\nabla \times \vec{H} = \vec{J} + j\omega\varepsilon\vec{E}$$

$$\vec{J} = \sigma\vec{E}$$

σ : conductivity of material medium ($\Omega^{-1}\text{m}^{-1}$)

$$\nabla \times \vec{H} = \sigma\vec{E} + j\omega\varepsilon\vec{E} = \vec{E}(\sigma + j\omega\varepsilon) = j\omega\varepsilon\left(1 + \frac{\sigma}{j\omega\varepsilon}\right)\vec{E}$$

since $\varepsilon \rightarrow \varepsilon\left(1 + \frac{\sigma}{j\omega\varepsilon}\right)$ **then** $\nabla^2 \vec{E} = -\omega^2\mu\varepsilon\left(1 + \frac{\sigma}{j\omega\varepsilon}\right)\vec{E}$

Wave in Material Medium

$$\nabla^2 \vec{E} = -\omega^2 \mu \epsilon \left(1 + \frac{\sigma}{j\omega\epsilon} \right) \vec{E} = \gamma^2 \vec{E}$$

$$\gamma^2 = -\omega^2 \mu \epsilon \left(1 + \frac{\sigma}{j\omega\epsilon} \right)$$

γ is complex propagation constant

$$\gamma = j\omega\sqrt{\mu\epsilon} \sqrt{1 + \frac{\sigma}{j\omega\epsilon}} = \alpha + j\beta$$

α : associated with attenuation of wave

β : associated with propagation of wave

Wave in Material Medium

Solution: $\vec{E} = \hat{x}E_o e^{-\gamma z} = \hat{x}E_o e^{-\alpha z} e^{-j\beta z}$ decaying exponential

$$\alpha = \omega \sqrt{\frac{\mu\epsilon}{2}} \left[\sqrt{1 + \left(\frac{\sigma}{\omega\epsilon} \right)^2} - 1 \right]^{1/2}$$

$$\beta = \omega \sqrt{\frac{\mu\epsilon}{2}} \left[\sqrt{1 + \left(\frac{\sigma}{\omega\epsilon} \right)^2} + 1 \right]^{1/2}$$

Complex intrinsic impedance

Magnetic field

$$\vec{H} = \hat{y} \frac{E_o}{\eta} e^{-\alpha z} e^{-j\beta z}$$

$$\eta = \sqrt{\frac{j\omega\mu}{\sigma + j\omega\epsilon}}$$

Wave in Material Medium

Phase Velocity: $v_p = \frac{\omega}{\beta} = \sqrt{\frac{2}{\mu\epsilon}} \left[\sqrt{1 + \left(\frac{\sigma}{\omega\epsilon} \right)^2} + 1 \right]^{-1/2}$

Wavelength: $\lambda = \frac{2\pi}{\beta} = \frac{\sqrt{2}}{f \sqrt{\mu\epsilon}} \left[\sqrt{1 + \left(\frac{\sigma}{\omega\epsilon} \right)^2} + 1 \right]^{-1/2}$

Special Cases

1. Perfect dielectric $\sigma = 0$ air, free space

$$\alpha = 0 \quad \text{and} \quad \beta = \omega \sqrt{\mu\epsilon}$$

Wave in Material Medium

2. Lossy dielectric

Loss tangent: $\frac{\sigma}{\omega\epsilon} \ll 1$

$$\alpha \approx \frac{\sigma}{2} \sqrt{\frac{\mu}{\epsilon}} \left[1 - \frac{\sigma^2}{8\omega^2\epsilon^2} \right] \approx \frac{\sigma}{2} \sqrt{\frac{\mu}{\epsilon}}$$

$$\beta \approx \omega \sqrt{\mu\epsilon} \left[1 + \frac{\sigma^2}{8\omega^2\epsilon^2} \right] \approx \omega \sqrt{\mu\epsilon}$$

$$\eta \approx \sqrt{\frac{\mu}{\epsilon}} \left[1 - \frac{3\sigma^2}{8\omega^2\epsilon^2} + j \frac{\sigma}{2\omega\epsilon} \right]$$

Wave in Material Medium

3. Good conductors

Loss tangent: $\frac{\sigma}{\omega\epsilon} \gg 1$

$$\gamma = \sqrt{j\omega\mu(\sigma + j\omega\epsilon)} \simeq \sqrt{j\omega\mu\sigma}$$

$$\alpha = \sqrt{\pi f \mu \sigma} \quad \beta = \sqrt{\pi f \mu \sigma}$$

$$\eta = \sqrt{\frac{j\omega\mu}{\sigma + j\omega\epsilon}} = \sqrt{\frac{\pi f \mu}{\sigma}} (1 + j)$$

Material Medium

	α attenuation	β propagation	η	d_p	H, E	Examples
PEC	∞	-	0	0	0	supercond
Good conductor	$\sqrt{\frac{\omega\mu\sigma}{2}}$	$\sqrt{\frac{\omega\mu\sigma}{2}}$	$\sqrt{\frac{\pi f \mu}{\sigma}}(1+j)$	$\frac{1}{\sqrt{\pi f \mu \sigma}}$	finite	copper
Poor conductor	$\frac{\sigma}{2} \sqrt{\frac{\mu}{\epsilon}}$	$\omega \sqrt{\mu \epsilon}$	$\frac{\sqrt{\mu / \epsilon}}{\left(1 - j \frac{\sigma}{2\omega\epsilon}\right)}$	$\frac{2}{\sigma} \sqrt{\frac{\epsilon}{\mu}}$	finite	Ice
Perfect dielectric	0	$\omega \sqrt{\mu \epsilon}$	$\sqrt{\mu / \epsilon}$	∞	finite	air

Radiation - Vector Potential

Assume time harmonicity $\sim e^{j\omega t}$

$$\nabla \times \vec{E} = -j\omega\mu\vec{H} \quad (1)$$

$$\nabla \times \vec{H} = \vec{J} + j\omega\varepsilon\vec{E} \quad (2)$$

$$\nabla \cdot \vec{D} = \rho / \varepsilon \quad (3)$$

$$\nabla \cdot \vec{B} = 0 \quad (4)$$

Radiation - Vector Potential

Using the property: $\nabla \cdot (\nabla \times \text{any vector}) = 0$

$$\nabla \cdot \vec{B} = 0 \Rightarrow \exists \vec{A} \text{ such that } \nabla \times \vec{A} = \vec{B}$$

$$\nabla \cdot (\nabla \times \vec{A}) = 0$$

$\vec{A}$: vector potential

$$\nabla \times \vec{E} = -j\omega \nabla \times \vec{A} \Rightarrow \nabla \times (\vec{E} + j\omega \vec{A}) = 0$$

Vector Potential

$$\nabla \times \text{vector} = 0 \Rightarrow \text{vector} = \nabla \psi$$

$$\left(\vec{E} + j\omega \vec{A} \right) = -\nabla \phi \text{ where } \phi \text{ is the scalar potential}$$

Since a vector is uniquely defined by its curl and its divergence, we can choose the divergence of $\vec{A}$

choose $\vec{A}$ such that

$$\nabla \cdot \vec{A} + j\omega\mu\epsilon\phi = 0 \leftarrow \text{Lorentz condition}$$

Vector Potential

$$\nabla \times \vec{B} = \mu \vec{J} + j\omega\mu\epsilon \vec{E}$$

$$\nabla \times \nabla \times \vec{A} = \mu \vec{J} + j\omega\mu\epsilon (-j\omega \vec{A} - \nabla \phi)$$

$$-\nabla^2 \vec{A} + \nabla (\nabla \cdot \vec{A}) = \mu \vec{J} + \omega^2 \mu\epsilon \vec{A} - j\omega\mu\epsilon \nabla \phi$$

$$-\nabla^2 \vec{A} + \nabla (-j\omega\mu\epsilon \phi) = \mu \vec{J} + \omega^2 \mu\epsilon \vec{A} - j\omega\mu\epsilon \nabla \phi$$

$$\nabla^2 \vec{A} + \omega^2 \mu\epsilon \vec{A} = -\mu \vec{J}$$

**D'Alembert's
equation**

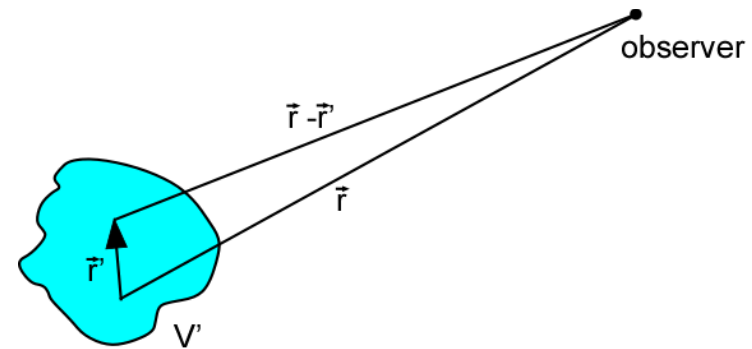
Vector Potential

Three-dimensional free-space Green's function

$$G(\vec{r}, \vec{r}') = \frac{e^{-j\beta|\vec{r}-\vec{r}'|}}{4\pi|\vec{r}-\vec{r}'|} \mu$$

Vector potential

$$\vec{A}(\vec{r}) = \mu \iiint_{V'} \frac{\vec{J}(\vec{r}') e^{-j\beta|\vec{r}-\vec{r}'|}}{4\pi|\vec{r}-\vec{r}'|} dv'$$



From $\vec{A}$, get $\vec{E}$ and $\vec{H}$ using Maxwell's equations

Magnetostatics

When $d/dt \rightarrow 0$ (or $\omega = 0$)

Poisson's Equation $\nabla^2 \vec{A} = -\mu \vec{J}$

The vector potential is
$$\vec{A}(\vec{r}) = \mu \iiint_{V'} \frac{\vec{J}(\vec{r}')}{4\pi |\vec{r} - \vec{r}'|} dv'$$

For a thin wire carrying current I , $\vec{J}dV' \rightarrow I d\vec{\ell}'$ along the wire,

$$\vec{A}(\vec{r}) = \frac{\mu I}{4\pi} \int_C \frac{d\vec{\ell}'}{|\vec{r} - \vec{r}'|}$$

Inductance

The flux is $\Phi = \oint_C \vec{A} \cdot d\vec{l} = \frac{\mu I}{4\pi} \oint_C \oint_C \frac{d\vec{l} \cdot d\vec{l}'}{R}$

where $R = |\vec{r} - \vec{r}'|$

Thus inductance is $L \triangleq \frac{\Phi}{I} = \frac{\mu}{4\pi} \oint_C \oint_C \frac{d\vec{l} \cdot d\vec{l}'}{R}$